

Powering SiPMs and front-end electronics in HEP detectors The ALDO2 ASIC

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SiPMs in particle physics experiments

- Silicon photomultipliers (SiPMs, aka MPPCs) are becoming widely adopted in many particle physics experiments at accelerators
 - CALICE, T2K, Belle II, CMS, GlueX, KLOE-2, LHCb, MEG II, NA62, ...
 - ... and more experiments are considering them for upgrades
- SiPM advantages are many and widely known
 - Excellent photon counting performance, high detection efficiency, high gain, fast response, compactness, immunity to magnetic field, low bias voltage, decreasing cost per unit of area, ...
- And so the drawbacks
 - High dark counts, moderate radiation hardness, performance strongly dependent on bias voltage and temperature, high capacitance per unit of area, crosstalk, after-pulses, ...
- All the experiments mentioned above adopt similar strategies to mitigate drawbacks
 - Operation at low and stable temperature
 - Precise voltage and current control
 - (Periodic annealing)







Powering SiPMs at HL-LHC

- At HL-LHC challenges with SiPMs will further increase
- Use case 1: CMS Barrel Timing Layer (LYSO bars + SiPMs)
 - 768 SiPMs per tray (with shared bias supply)
 - About 1 mA/SiPM at EOL irradiation level \rightarrow 0.8 A total
 - Up to 40 m distance between bias PSU and SiPMs
 - No access to the detector after the tracker is mounted
 - Limited power budget, bias voltage decreased as dark current increases
- Use case 2: CMS High-Granularity Calorimeter (plastic scintillator tiles + SiPMs)
 - Similar needs









ALDO2 ASIC = Adjustable Low Drop-Out linear regulator with 2 functions

- SiPM bias voltage regulation, protection, and monitoring
 - Fine segmentation
 - 768 channels for each HV bias supply
 - \rightarrow 16 channels for ALDO HV channel
 - Independent setting and switch-off of each SiPM group
 - ~Few V bias voltage adjustment with ~tens mV resolution
 - Dark current measurement for each group
- FE ASIC power supply regulation
 - Filter noise and transients on the power supply
 - Enhance thermal and load stability
 - Fine segmentation
 - 6 FE ASIC for each bPOL DCDC
 → 1 FE ASIC for each ALDO
 - Generate aux voltages (e.g. GBT-SCA 1.5V)

SiPM bias voltage regulation and protection (HV)



FE ASIC power supply regulation (LV)



LV part

- Input voltage: 1.6 V to 3.3 V
- Output voltage: adjustable (min 0.6 V)
- Output current: 0.6 A
- Minimum dropout: 0.4 V
- PSRR/line regulation: -40 dB (10 μV/mV) at max load
- Noise: 30 μV RMS (10 Hz 100 MHz)
- Thermal stability: 50 ppm/°C
- Thermal overload and over-current protection (0.9 A limit)
- Shutdown capability

HV part

- 2 independent channels
- Output voltage: adjustable, 10-60 V
- Output current: 32 mA
- Minimum dropout: 1 V at full load
- PSRR/line regulation: -40 dB (10 mV/V) at max load
- Noise: 350 μV RMS (10 Hz 100 MHz)
- Thermal overload and over-current protection (50 mA limit)
- Shutdown capability
- Output current measurement

Technology choice

- onsemi 0.35 μm I3T80 CMOS technology
 - Qualified for radiation hardness by CERN DCDC group (F. Faccio et al., DOI: 10.1109/TNS.2010.2049584), used by FEAST DCDC and some bPOL DCDCs
 - Has all the devices needed (70 V N- and P-channel DMOSFETs, standard LV MOSFETs, etc.)
- Still some critical points for our application
 - ELT required for LV NMOS
 - HV NMOS requires custom layout to reduce leakage above 100 krad
 - On resistance of HV MOSFETs increases with displacement damage
 - Not super critical for our application, use much larger transistors
 - Use all the standard RHBD techniques (guard rings, spacing, etc.)
 - HV transistors have thin-oxide gates, max V_{gs} limited to 3.6 V









- Standard LDO topology (PMOS output MOS)
 - >500 mA output current, very large output MOS
 - Adjustable gain with external resistors
 - External compensation with low ESR capacitor
- Embedded bandgap
 - 3 bandgaps included (MOS, PNP, NPN based)
 - MOS-based selected due to best radiation hardness
- Output can be disabled to power cycle the load
- Overcurrent protection (OCP) and overtemperature protection (OTP) protect the device in case of load issues (shorts, etc.)





- LDO topology (HV PMOS output MOS), driven by requirements
 - Improve stability and noise of bias supply → active feedback
 - Capability to disable each channel independently in case of issues in one SiPM array → series switch required
 - − Avoid parasitics on FE ASIC input node \rightarrow high-side regulation
 - − Output current measurement \rightarrow current mirroring
 - High efficiency → PMOS output HV MOS for low dropout (e.g. 45 V input, 42 V output, 93% efficiency)
- Output voltage adjustable in two ways
 - Gain resistors to match SiPM model needs
 - Voltage reference using a DAC outside of the chip (FE ASIC in BTL or GBT-SCA in HGCAL)

Main challenges of this architecture

- Driving LV (thin-oxide) gate of HV MOS
- Gain is high (30-50 V/V)
 - DAC noise and input offsets must be minimized
- External compensation with low ESR tantalum capacitors
 - Few options rated 50 V, rather bulky (7.3x4.3x1.8 mm³)





HV part – Error amplifier

- The error amplifier has to operate between HV rail (up to 60 V) and GND
 - Mix of HV and LV transistors + bias lines to limit DC voltage excursion within LV MOS max ratings
 - Bias lines (BIASH and BIASL) are provided externally (ALDO2v0-v1) or generated internally (ALDO2v2)
- High loop gain (70 dB) to minimize regulation errors
- The regulator is disabled by acting on the error amplifier



HV part – Output current protection and monitoring

- The over-current protection and output current measurement use a similar strategy
 - A fraction of the output current is mirrored in the two circuits
 - The over-current circuit has a constant current limit (no foldback)
 - The current measurement circuit further mirrors the output current with tunable ratio (1/40 and 1/800)
 - An external resistor converts the output current to a voltage, matching the ADC full-range and SiPM specs



Over-temperature protection

- The over-temperature protection acts both on HV and LV part
- Internal PTC voltage compared to an external threshold
- PTC element based on weak-inversion composite MOS transistor (10.1093/ietele/e91-c.4.662, 10.5772/39231 and 10.1049/el.2010.1346)



Parameter		Test Conditions	Min	Тур	Max	Unit
V _T	Temperature sensor voltage	Temp. = 27 °C	163.9	166.4	168.7	mV
%V _T	Relative error		-1.5		1.4	%
dV _T /dT	Thermal coefficient		0.411	0.424	0.437	mV/°C
dV_T/dV_{DD}	Power supply coefficient			6.8		mV/V
dV _T /dI _D	Bias current coefficient	Linearized around 6 µA		1		mV/μA

Prototyping

- 2019: ALDO2v0
- 2021: ALDO2v1
 - Re-centered bandgap trimming, higher default current limits
- 2022: ALDO2v2 (final production)
 - Self-bias, extend input voltage range
 - 45k chips produced



Die area 2.47 x 1.95 mm² Package QFN64





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Selected measurements



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Radiation hardness

Irradiation performed in several facilities

- TID: KIT (ambient T) and CERN (cold T)
- Neutrons: LENA and JSI
- Heavy ions: LNL

Radiation levels at 3000 fb ⁻¹ in CMS BTL			
TID	Neutron fluence	Ch. hadron fluence	
3.2 Mrad	1.9e14 n _{eq} cm ⁻²	1.5e13 cm ⁻²	

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- Typical thermal stability before irradiation 50 ppm/°C
- Irradiation does not significantly affect thermal stability

Radiation hardness – Heavy ions

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- Single event transients (SET) observed
- Cause: OTP circuitry (comparator)
- Mitigation: more filtering of OTP threshold
- Extrapolation with conservatively low threshold LET (1 MeV cm² mg⁻¹) gives a SET rate of 10 mHz in whole BTL
- Further mitigation: FE ASIC can disable OTP circuitry by software

Take-home-message with SET: Be careful with high-gain, highbandwidth and open-loop circuits

SET cross section



Line regulation – post-irradiation

- Radiation damage degrades line regulation
- LV part
 - Most critical because bPOL output is fixed during operation
 - At 7 Mrad and full load (520 mA), line regulation starts degrading when input voltage gets below 1.7 V (500 mV dropout)
 - Final choice for BTL: 600 mV dropout (1.8 V ALDO input voltage)
- HV part
 - Less critical because dropout can be adjusted on PSU
 - Min dropout: 2-3 V at detector EOL, >90-95% efficiency
 - Large degradation in HV PMOS R_{on} at 1e15 n_{eq} cm⁻² (x5 above EOL level)







Yield and testing

- 552 ALDO2v2 measured in a test socket
 - Samples picked randomly from the full production
- LV part: 100% yield
- HV part:
 - 99.5% yield when counting partially working chips (e.g. with one channel that is always enabled)
 - 99.1% yield when counting only fully working chips
- Final decision: no individual chip testing, chips tested on FE PCB (with 2 ALDOs, exp yield >98%)





- We have developed ALDO2, a multi-function rad-hard ASIC for the power management of SiPM-based HEP experiments
- Performance was validated over the course of 4 years of development, laboratory measurements and detector integration
- Radiation hardness was fully qualified up to 20 Mrad, 1e15 n_{eq} cm⁻² and heavy ions
- Full production has been completed, 45k chips produced
- ALDO2 is ready to be installed in two CMS detectors, BTL and HGCAL

Future prospects

- A possible ALDO3 would likely include DAC for bias adjustment, ADC for monitoring, and digital logic
- onsemi I3T80 fab2 (used for ALDO2) closed in 2022
 - I3T80 still available in US fab, not yet qualified for radiation
 - Availability through Europractice uncertain
- Alternative rad-hard HV technologies could be investigated





Thanks for your attention



Backup





 The LV part is completely independent from the HV part (this is the case when used with GBT-SCA)







- Three bandgaps included (only one selected in final application)
 - PNP-based
 - NPN-based
 - MOS-based
- The reason for this is that the expected radiation levels are right on the edge where bipolar-based bandgaps could still be used (and they offer superior stability)
- MOS-based bandgap can be trimmed to compensate corner spread and/or simulation mismatch due to the use of ELTs
 - «Passive» trimming: acts directly on bandgap resistors (resistors can be shorted externally to slightly tune the gain of the bandgap)
 - 4 bits, ~8 mV LSB (± ~10% trimming wrt 630 mV)
 - The trimming strategy relies on the fact that trimming can be done on a wafer-by-wafer basis





- In addition to the main regulator (0.5 A) there is also an auxiliary one (max 20 mA)
 - This was originally intended to provide an external voltage reference to the front-end ASIC
 - It is also used to generate the voltage reference for the over-temperature protection
- The design of the LV error amplifiers is quite standard
 - Typical closed-loop gain: 1-2
 - Loop gain: 54 dB
 - Single-stage "mirrored" amp, NMOS input
- Output voltage is adjustable by changing feedback resistors
- Any bandgap can be selected as input of the voltage reference
- Compensation on output node with low ESR tantalum capacitors
- Overcurrent protection is also very similar to HV part

HV part – Power domains

- The HV part is completely independent of the LV part
- Aux circuitry of the HV part has its own LV power domains
- No power-on sequence is required





HV part – Error amplifier (2)

- The regulator can be disabled (default off)
- Bias current is provided externally (not from bandgap) to keep LV and HV independent



HV part – Overcurrent protection

- The overcurrent circuit compares the output current to a gate threshold
- The error amplifier is disabled if current gets too high
- Current limit is constant (no foldback, operates together with over-temperature)



HV part – Output current measurement

- The current measurement circuit further mirrors the output current
- The ratio is tunable (1/40 and 1/800), adapting to BOL and EOL conditions
- An external resistor converts the output current to a voltage, matching the ADC full-range and SiPM specs



Tantalum capacitors

- For the stability, both HV and LV regulators rely on low ESR tantalum capacitors on output nodes
- Radiation hardness is not a problem
 - Used widely in space applications due to they higher reliability and compact size
 - Sensitive to humidity
 - No problem in HEP detectors where ambient is controlled
 - Tested up to 5 Mrad (https://core.ac.uk/download/pdf/227725417.pdf)
- Several models available for LV in small packages (22-100 μF, ESR < 200 mohm)
- Less models for HV ones
 - 50 V/63 V capacitors are thick
 - Selected model: KEMET T521V106M050ATE090 (50 V, 10 μF, 90 mohm ESR, height 1.8 mm)
 - Costly

Accuracy of I-V curves

 The I-V curve with no load (only feedback resistors) shows that the accuracy of the current measurement circuitry is as expected from simulations and previous measurements (<2% spread)





- Spread from the whole production (ALDO2v2) is small, no need for bandgap trimming
 - 0.7% RMS on bandgap/LV regulator





Blue: ALDO2v2 from same wafer

- <0.1% RMS on HV regulator</p>





Red:

ALDO2v2 sampled randomly from 6 wafers

Accelerated aging

- Accelerated aging using Arrhenius equation gives x2 acceleration factor every 10 °C increase
- 10 years of operation can be accelerated to 1 month if temperature is set 70 °C above operating temp
- In BTL operating temperature will be -30 °C

- We decided to perform accelerated aging at 80 °C
- Input voltages: 1.8 V and 45 V, output voltages: 1.2 V and 42 V (40 V), nominal load currents: 500 mA and 30+30 mA
- Run fine for >700 hours (30 days), with 1 chip
- Temperature with on-board thermometer: 88 °C



Operation at extremely low temperature

- We tried to do the same thing at low temperature (climatic chamber at -60 °C and -70 °C)
- Run fine for 240 hours (10 days), with 1 chip
- Temperature with on-board thermometer: -52 °C or -63 °C



Operation beyond specs

- What happens if the chip is operated beyond specs?
- We increased the temperature to 90 °C, increased input voltages to 2.5 V and 50 V. Now BIASL voltage is at 3.5 V (beyond specs of diode-connected LV MOS and possibly beyond SOA of DMOS)
- Both regulators worked for 10 additional days



Biasing diodes – IV curves

- After irradiation, we measure the diode IV curves using a Keithley 4200 semiconductor analyzer at different temperatures (-50 °C to +70 °C)
- NMOS-based diode (diode 'L') has small drift due to radiation
 - Up to 200 mV voltage drift at fixed current
- PMOS-based diode (diode 'H') has significant drift
 - Up to 700 mV voltage drift at fixed current



I3T80 Fab2 closure

- I3T80 Fab2 in Oudenaarde (Belgium) was closed in June 2022
- There is a backup solution to Fab2: I3T80 fab in Gresham (USA)
 - 8-inch wafers
 - Different rules for metal slotting
 - Radiation hardness unknown
- In 2021 we did an MPW of ALDO2v1 in Gresham fab
 - Dies are unpackaged
 - Available for radiation hardness qualification
- Future MPW runs with Europractice uncertain

TID online monitoring - LV



- LV regulator is fully operational up to 20 Mrad at ambient temperature and 520 mA load
- All three bandgaps are well within a few % range at 3.2 Mrad (PNP and MOS ones are better, <1%)

TID online monitoring - HV



- HV regulator is fully operational up to 20 Mrad at ambient temperature and 20 mA load
- Channel A started working in current-limited mode at about 12 Mrad (marker:), lacause the over-current limit lowered during irradiation. Recovered temporarily by switching off the over-current protection

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05/09/2023

39

Bandgap voltage references

- Bandgaps drifted by few percent after irradiations. All are well within usable range at nominal radiation levels
 - Best: MOS bandgap, Second best: PNP bandgap



Value

1.22 V

1.22 V

1.27 V

1.15 V

1.07 V

Var

1%

4%

-6%

-12%

T Drift

-100

ppm/°

-270

ppm/°

ppm/°(

ppm/°

NPN

Pre-irrad

2.5e14 cm⁻²

1e15 cm⁻²

7 Mrad

20 Mrad



;	PNP	Value	Var	T Drift
2	Pre-irrad	1.17 V	-	-20 ppm/°C
	2.5e14 cm ⁻²	1.21 V	3-4%	-
C	1e15 cm ⁻²	1.26 V	8%	150 ppm/°C
C	7 Mrad	1.2 V	2%	-10 ppm/°C
0	20 Mrad	1.21 V	4%	-40 ppm/°C



MOS	Value	Var	T Drift
Pre-irrad	625 mV	-	70 ppm/°C
2.5e14 cm ⁻²	625 mV	0%	-
1e15 cm ⁻²	621 mV	<1%	120 ppm/°C
7 Mrad	637 mV	2%	70 ppm/°C
20 Mrad	669 mV	7%	100 ppm/°C

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